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#### <span id="page-0-0"></span>EEN452 - Control and Operation of Electric Power Systems Part 6: Power system stability fundamentals <https://sps.cut.ac.cy/courses/een452/>

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After this part of the lecture and additional reading, you should be able to . . .

- **<sup>1</sup>** . . . understand the basic classifications of power system stability;
- **<sup>2</sup>** . . . be able to identify and perform stability analysis problems; and,
- **<sup>3</sup>** . . . propose methods for stabilizing power systems.



# <span id="page-3-0"></span>**[Introduction](#page-3-0)**

- **[Voltage Stability](#page-11-0)**
- **[Rotor Angle Stability](#page-25-0)**

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# **[Voltage Stability](#page-11-0)**

### **[Rotor Angle Stability](#page-25-0)**

### **[References](#page-58-0)**

## **2 Voltage stability fundamentals**

<span id="page-12-0"></span>



$$
P_l = \frac{U_l U_N}{X_e} \sin \phi \qquad Q_l = \frac{U_l U_N \cos \phi - U_l^2}{X_e}
$$

$$
P_l^2 + \left(Q_l + \frac{U_l^2}{X_e}\right)^2 = \left(\frac{U_l U_N}{X_e}\right)^2 \Rightarrow \left(U_l^2\right)^2 + \left(2Q_l X_e - U_N^2\right)U_l^2 + X_e^2(P_l^2 + Q_l^2) = 0
$$
\n(2.1)

### **2 Voltage stability fundamentals**



To have (at least) one solution:

$$
\left(2Q_lX_e - U_N^2\right)^2 - 4X_e^2\left(P_l^2 + Q_l^2\right) \ge 0 \Rightarrow -\left(\frac{P_lX_e}{U_N}\right)^2 - \frac{Q_lX_e}{U_N^2} + 0.25 \ge 0
$$







- any *P<sup>l</sup>* can be reached provided *Ql* is adjusted (but *U<sup>l</sup>* may be unacceptable)
- dissymmetry between *P<sup>l</sup>* and *Q<sup>l</sup>* due to reactive transmission impedance
- locus symmetric w.r.t. *Q<sup>l</sup>* axis; this does no longer hold when transmission resistance is included

Under a constant load power factor  $\cos \phi$  (i.e.,  $Q_l = P_l \tan \phi$ ), we get from Eq. [\(2.1\)](#page-12-0):

$$
P_l^2 + \frac{U_N^2}{X_e} \tan \phi P_l - \frac{U_N^4}{4X_e^2} = 0
$$

Then, we get the maximum power:

$$
P_l^{max} = \frac{\cos\phi}{1+\sin\phi}\frac{U_N^2}{2X_e} \qquad Q_l^{max} = \frac{\sin\phi}{1+\sin\phi}\frac{U_N^2}{2X_e} \qquad U_l^{max} = \frac{U_N}{\sqrt{2}\sqrt{1+\sin\phi}}
$$

Or, for the extreme cases:

$$
\cos \phi = 1: \qquad P_l^{max} = \frac{U_N^2}{2X_e} \qquad Q_l^{max} = 0 \qquad U_l^{max} = \frac{U_N}{\sqrt{2}}
$$
\n
$$
\cos \phi = 0: \qquad P_l^{max} = 0 \qquad Q_l^{max} = \frac{U_N^2}{4X_e} \qquad U_l^{max} = \frac{U_N}{2}
$$

 $\sim$ 







- $\bullet$  for given power  $(P_i)$ 
	- 1 solution with "high" voltage and "low" current (normal operating point)
	- 1 solution with "low" voltage and "high" current
- compensating the load increases the maximum power but the "critical" voltage approaches normal values
- curves that provide similar information:
	- *Q* − *V* or *S* − *V* under constant tan  $\phi$ ,  $Q - V$  under constant  $P$ , etc.





- o in real systems, much more complicated
	- no infinite bus, voltage control by generators (AVR)
	- multiple loads and generators
	- complex, meshed transmission system with resistive components as well
	- voltage sensitive loads and restorative behavior
	- etc.





- Pre-fault loadability curve of the system  $(\Sigma_1)$
- Fault occurs in the system
	- **Loadability curve is shrunk to**  $\Sigma$ **<sub>2</sub>**
	- **<sup>2</sup>** Post-fault consumption is decreased due to depressed voltages (voltage sensitive loads). If point *B* is outside  $\Sigma_2$  then there is no solution and we have *short-term voltage instability*
- Loads try to restore consumption to pre-fault point *A* now outside the loadability curve
- *Long-term instability* leading to voltage collapse





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### **2 Voltage instability: example (RAMSES with Nordic system)**







- Series compensation: very effective but expensive
- Shunt compensation: cheapest mechanism
- SVC and STATCOM devices
- Adjustment of generator active power productions
- Adjustment of generator voltages
- Block load restoration (e.g., through load tap changers) : effective but sometimes too slow
- Undervoltage load shedding : effective but expensive, last resort







### <span id="page-25-0"></span>**<sup>1</sup> [Introduction](#page-3-0)**

# **<sup>2</sup> [Voltage Stability](#page-11-0)**

### **<sup>3</sup> [Rotor Angle Stability](#page-25-0)**

- [Transient stability](#page-27-0)
- [Small-disturbance angle stability](#page-34-0)
- [Summary](#page-56-0)

### **<sup>4</sup> [References](#page-58-0)**

- most of the electrical energy today is generated by synchronous machines
- o in normal system operation:
	- all synchronous machines rotate at the same electrical speed  $\omega_0 = 2\pi f_n$
	- the mechanical and electromagnetic torques acting on the rotating masses of each generator balance each other

 $\frac{20}{2H_i}(T_{mi}-T_{ei})$ 

the phase angle differences between the internal e.m.f.'s of the various machines are constant (synchronism)

**following a disturbance, there is an imbalance between the two torques and the rotor speed varies**

 $\dot{\omega}_i = \frac{\omega_0}{2D}$ 

**rotor angle stability deals with the ability to keep/regain synchronism after being subject to a disturbance**





- <span id="page-27-0"></span>Transient (angle) stability deals with the ability of the system to keep synchronism after being subject to a **large** disturbance
- typical "large" disturbances:
	- short-circuit cleared by opening of circuit breakers
	- more complex sequences: backup protections, line autoreclosing, etc.
- $\bullet$  the nonlinear behavior of the generator and its controllers must be taken into account
	- numerical integration of the differential-algebraic equations is used
- unacceptable consequences of transient instability:
	- generators tripped due to loss of synchronism (to avoid equipment damages)
	- long-lasting voltage dips created by large angle swings (disturb customers)

### **3.1 Transient (angle) stability**





<span id="page-28-0"></span>where in steady-state  $P_{m0} = P_{e,max} \sin \theta_0$ 



Multiplying both sides of Eq. [\(3.1\)](#page-28-0) with  $\dot{\theta}$  and integrating:

$$
\frac{1}{2}M\dot{\theta}^2-\int\limits_0^t\left(P_{m0}-P(\theta)\right)\dot{\theta}dt=C
$$

Changing the integration variable  $(x = \theta(t))$ 

$$
\frac{1}{2}M\dot{\theta}^2+\int\limits_{\theta_0}^{\theta}(P(x)-P_{m0})\,dx=C
$$

"kinetic" energy + "potential" energy = Constant







 $\theta$ <sub>e</sub>

*The system is stable if there exists an angle* θ *i <sup>p</sup> such that the areas are equal*  $(A_{acc} = A_{dec})$ 

Or, for a given  $\theta_e$ ,  $A_{acc} - A_{dec} < 0$ 

$$
A_{acc} = \int\limits_{\theta_u^0}^{\tilde{}} (P_d(x) - P_{m0}) dx
$$

$$
A_{dec} = \int\limits_{\theta_e}^{\theta_p^i} (P_p(x) - P_{m0}) dx
$$





### *The system is stable if there exists an angle* θ *i <sup>p</sup> such that the areas are equal*  $(A_{acc} = A_{dec})$

Or, for a given  $\theta_e$ ,  $A_{acc} - A_{dec} < 0$ 

### **Curves:**

- *during fault*: capability of evacuating power on the network decreased due to low voltages
- *post-fault*: system weaker owing to the fault clearing actions (e.g., line tripping)

### **Critical clearing time (** $t_e = t_c$ **):**

- Maximum fault duration so that the system returns to equilibrium
- When the system is at the stability limit :  $A_{acc} - A_{dec} = 0$ and  $\theta_e = \theta_c = \theta(t_c)$

### **3.1 Transient (angle) stability: example (RAMSES with 5-bus system)**





#### **Disturbances:**

- 6-cycle (120ms) short-circuit without impedance on line "1-3", next to bus 3, cleared by opening that line, when the generator produces 450 MW
- the same fault cleared without line opening, when the generator produces 450 MW
- $\circ$  the same sequence as above, but with the generator producing 400 MW



- Modifying the pre-disturbance operating point:
	- reducing the active power generation
	- operating with higher excitation
- Automatic emergency controls:
	- actions on network: line auto-reclosing, fast series capacitor reinsertion, fast fault clearing - single pole breaker operation
	- actions in generators: (turbine) fast valving, generation shedding
	- action on "load": dynamic braking
- Other means:
	- $\bullet$  equip generators with fast excitation system
	- control voltage at intermediate points in a long corridor: through synchronous condensers or static var compensators.

# <span id="page-34-0"></span>**3.2 Small-disturbance angle stability**

- Small-signal (or small-disturbance) angle stability deals with the ability of the system to **keep synchronism** after being subject to a "small disturbance"
- "small disturbances" are those for which the system equations can be **linearized** around an equilibrium point
	- tools from linear system theory can be used (in particular eigenvalue and eigenvector analysis)
- $\bullet$  following a small disturbance, the variation in electromagnetic torque  $T_e$ can be decomposed into:

 $\triangle T_e = K_s \triangle \delta + K_d \triangle \omega$ 

 $K_s \triangle \delta$ : synchronizing torque  $K_d \triangle \omega$ : damping torque

- **a decrease in synchronizing torque will eventually lead to aperiodic instability (machine "going out of step")**
- **a decrease in damping torque will eventually lead to oscillatory instability (growing oscillations)**





- A small "nudge" to the system (1 ms fault at bus 7) to excite the interarea modes.
- Oscillation of machines 1 and 2 against machines 3 and 4
- Period ∼ 2 s





### **Local modes (involve a small part of the system)**

- rotor angle oscillations of a single generator or a single plant against the rest of the system: *local plant mode*
	- can be studied using a one-machine infinite-bus system
- oscillations between rotors of a few generators close to each other: *inter-machine or inter-plant mode oscillations*
- typical range of frequencies of local plant and inter-plant modes: *0.7 to 2 Hz*
- may also be associated with inappropriate tuning of a control equipment (excitation system, HVDC converter, SVC, etc.): *control mode*



### **Global modes (involve large areas of the system, widespread effects)**

- $\bullet$  oscillations of a large group of generators in one area swinging against a group of generators in another area: *interarea mode*
- usually, the larger the group of generators, the slower the oscillations
- typical range of frequencies of interarea modes: *0.1 to 0.7 Hz*
- more complex to analyze and to damp

Let's consider an autonomous system described by the differential equations:  $\dot{x} = f(x)$ 

and 
$$
x^*
$$
 is an equilibrium point:  $f(x^*) = 0$ 

• If we linearize the system around the operating point and ignore higher order terms:

$$
\Delta \mathbf{x} = \dot{\mathbf{x}} = \left. \frac{\partial \mathbf{f}}{\partial \mathbf{x}} \right|_{\mathbf{x} = \mathbf{x}^*} \mathbf{x} = \mathbf{A} \mathbf{x}
$$

where  $\frac{\partial \bm{f}}{\partial \bm{x}}$  is the Jacobian of  $\bm{f}$  with respect to  $\bm{x}$ , and  $\bm{A} = \frac{\partial \bm{f}}{\partial \bm{x}}$ ∂*x x*=*x*∗ is the state matrix of the linearized system.





- $\bullet$  Let  $\lambda$  be a real eigenvalue of matrix **A** :
	- $\lambda$  < 0: The corresponding mode is stable (decaying exponential).
	- $\lambda > 0$ : The corresponding mode is unstable (growing exponential).
	- $\lambda = 0$ : The corresponding mode has integrating characteristics.
- $\bullet$  Let  $\lambda_{1,2} = \sigma \pm i\omega$  be a complex conjugate pair of eigenvalues of **A** :  $\mathbb{R}(\lambda_1, \lambda_2)$  < 0: The corresponding mode is stable (decaying oscillation).
	- $\mathbb{R}(\lambda_1, \lambda_2) > 0$ : The corresponding mode is unstable (growing oscillation).
	- $\mathcal{R}(\lambda_{1,2}) = 0$ : The corresponding mode is critically stable (undamped oscillation).
	- The following dynamic properties can be established:

\n- Oscillation frequency: 
$$
f = \frac{\omega}{2\pi}
$$
\n- Damping ratio:  $\zeta = \frac{-\sigma}{\sqrt{\sigma^2 + \omega^2}}$
\n

### **3.2 Dynamic Analysis of the Heffron-Phillips Model**







### **SMIB with classical generator model (mechanical damping torque**  $K_D = 0$



**Eigenvalues, synchronizing and damping torque coefficients**



**Eigenvalues on imaginary axis → system is critically stable**



### **SMIB including field circuit dynamics**



### **Eigenvalues, synchronizing and damping torque coefficients**



**Eigenvalues moved to the left because field circuit adds damping torque**



### **SMIB including excitation system**



### **Eigenvalues, synchronizing and damping torque coefficients**



### **Eigenvalues moved to the right by the excitation system**  $→$  **System is unstable!**

Let's assume again the linearized system with the state matrix  $\boldsymbol{A}$  ( $n \times n$ ) and  $\lambda_i$  is one of its non-zero eigenvalues. Then:

 $v_i$  is the right eigenvector of  $\lambda_i$ :

• 
$$
w_i
$$
 is the left eigenvector of  $\lambda_i$ :

$$
\bullet
$$
 In matrix form:

\n- $$
w_n^T
$$
  $\bigcup$
\n- It can be shown that:  $W = V^{-1}$  and  $WAV = \text{diag}(\lambda_i) = \Lambda$
\n

 $V = [v_1 ... v_n]$  *W* =

$$
f_{\rm{max}}
$$

$$
Av_i = \lambda_i v_i
$$

$$
\mathbf{A}^T \mathbf{w}_i = \lambda_i \mathbf{w}_i
$$

 $\sqrt{ }$ 

1

 $\begin{array}{c} \n\downarrow \\ \n\downarrow \n\end{array}$ 

 $\begin{array}{c} \begin{array}{c} \begin{array}{c} \begin{array}{c} \end{array} \\ \end{array} \end{array} \end{array}$  $w_1^7$ <br> $\vdots$  Now, consider a system with state vector *x*, input vector *u* and a scalar output *z*:

$$
\dot{x} = Ax + Bu
$$

$$
z = Cx + Du
$$

 $\bullet$  We change the variables ( $\mathbf{y} = \mathbf{W}\mathbf{x}$ )

$$
\dot{y} = WAVy + WBu = Ay + WBu
$$
  

$$
z = CVy + Du
$$

• For the *i*-th "mode"  $\lambda_i$  of the state matrix **A**: the larger  $(\bm{WB})_i = \bm{w}_i^T\bm{B},$  the more the mode can be controlled by  $\bm{u}$ the larger  $(\bm{C}\bm{V})_i=\bm{C}\bm{v}_i$  , the more the mode can be observed in  $\bm{z}$ 



### **3.2 Transfer function and residues**

We can now build the transfer function of the system:

$$
H(s) = \frac{Z(s)}{U(s)}
$$
  
=  $\mathbf{C}\mathbf{V}(s\mathbf{I} - \mathbf{\Lambda})^{-1}\mathbf{W}\mathbf{B} + \mathbf{D}$   
=  $[\mathbf{C}\mathbf{v}_1 ... \mathbf{C}\mathbf{v}_n]$  diag $(\frac{1}{s - \lambda_i})$   $\begin{bmatrix} \mathbf{w}_1^T \mathbf{B} \\ \vdots \\ \mathbf{w}_n^T \mathbf{B} \end{bmatrix} + \mathbf{D}$   
=  $\sum_{i=1}^n \frac{\mathbf{C}\mathbf{v}_i \mathbf{w}_i^T \mathbf{B}}{s - \lambda_i} + \mathbf{D} = \sum_{i=1}^n \frac{R_i}{s - \lambda_i} + \mathbf{D}$ 

The residue  $R_i$  relative to the *i*-th mode  $\lambda_i$ :

depends on both the observability and the controllability of the mode

would be zero in case of exact zero-pole cancellation

# **3.2 Synthesis of a stabilizing feedback using residues**





Consider a compensator using *z* as input and acting on *u*:

Which condition should be satisfied by the transfer function of the PSS in order to stabilize the critical mode  $\lambda_c$  of the uncompensated system?

The closed-loop transfer function is:

*H*(*s*) 1 − *KPSSH*(*s*)*G*1(*s*)*G*2(*s*)

If  $\tilde{s}$  is one of the closed-loop poles:

$$
1 - K_{PSS}H(\tilde{s})G_1(\tilde{s})G_2(\tilde{s}) = 0 \Leftrightarrow 1 - K_{PSS}\left[\sum_{i=1}^n \frac{R_i}{\tilde{s} - \lambda_i} + \mathbf{D}\right]G_1(\tilde{s})G_2(\tilde{s}) = 0
$$

# **3.2 Synthesis of a stabilizing feedback using residues**





Consider a compensator using *z* as input and acting on *u*:

Which condition should be satisfied by the transfer function of the PSS in order to stabilize the critical mode  $\lambda_c$  of the uncompensated system?

The closed-loop transfer function is:

$$
\frac{H(s)}{1-K_{\mathit{PSS}}H(s)G_1(s)G_2(s)}
$$

If  $\tilde{s}$  is one of the closed-loop poles:

$$
1 - K_{PSS}H(\tilde{s})G_1(\tilde{s})G_2(\tilde{s}) = 0 \Leftrightarrow 1 - K_{PSS}\left[\sum_{i=1}^n \frac{R_i}{\tilde{s} - \lambda_i} + \boldsymbol{D}\right]G_1(\tilde{s})G_2(\tilde{s}) = 0
$$

# **3.2 Synthesis of a stabilizing feedback using residues**





Let's consider a closed-loop pole  $\tilde{s}$  lying on the branch of the root locus which starts from the open-loop pole  $\lambda_c$ . When the compensator gain  $K_{PSS}$  tends to zero,  $\tilde{s}$  tends to  $\lambda_c$ .

○ Keeping only the dominant terms:

$$
1-R_cG_1(\lambda_c)G_2(\lambda_c)\lim_{K_{PSS}\to 0}\frac{K_{PSS}}{\tilde{s}-\lambda_c}=0 \Leftrightarrow R_cG_1(\lambda_c)G_2(\lambda_c)=\lim_{K_{PSS}\to 0}\frac{\tilde{s}-\lambda_c}{K_{PSS}}
$$

- In the complex plane  $\lim_{K_{PSS}\to 0}$  $\mathbf{\tilde{s}} - \lambda_c$  $\frac{1}{K_{PSS}}$  is a vector tangent to the branch of the root locus starting from λ*<sup>c</sup>* .
- In order to shift the eigenvalue to the left:
	- $\bullet$  the branch of the root locus should leave  $\lambda_c$  at an angle of 180 degrees. Thus,  $G_1(\lambda_c)G_2(\lambda_c)$  must be such that  $\angle G_1(i\omega_c)G_2(i\omega_c) = \pm 180 - \angle R_c$
	- $\bullet$  *R<sub>c</sub>* $G_1(\lambda_c)G_2(\lambda_c)$  should be a real negative number



### **Purpose:**

Provide additional *damping torque* component in order to prevent the system from becoming unstable.

### **Approach:**

Insert a feedback between *angular frequency* and *voltage setpoint* to "stabilize" a critical mode  $\lambda_c$ .

### **Block diagram:**



**Figure:** Block diagram of a simple PSS



### **Phase Compensation** *G*1**:**

 $\bullet$  shifts  $\lambda_c$  to the left in the complex plane by bringing a phase compensation according to the residue method :

 $∠G_1(\lambda_c) ≈ ∠G_1(iω_c) = ±180 − ∠R_c$ 

- *G*<sub>1</sub>(*s*)corresponds to one or several lead-lag filters
- $\bullet$  the latter are "tuned" to provide their maximum phase shift  $\phi_m$  at the frequency ω*<sup>c</sup>*

### **Washout Filter**  $G_2$ :

- in steady state and for slow variations, the PSS must not affect voltage regulation
- $\circ$   $G_2(s)$  is a high-pass filter
- $\bullet$   $T_w$  is taken large enough to not modify the phase angle of  $G_1(s)$  for frequencies around ω*<sup>c</sup>* . For instance:

$$
\frac{10}{T_w}\backsimeq \frac{\omega_c}{10}
$$



### **Gain** *KPSS***:**

• adjusted until the corrected mode  $\tilde{\lambda}_c$  has a damping ratio :

$$
\zeta = \frac{-\Re(\tilde{\lambda}_c)}{\sqrt{\Re(\tilde{\lambda}_c)^2+\Im(\tilde{\lambda}_c)^2}} \geq 0.05-0.10
$$

- $\bullet$  while  $K_{PSS}$  is increased, the other eigenvalues are monitored since they might move to the right (*the residue method allows controlling a single mode*)
- $\bullet$  for excessive values of  $K_{PSS}$ , the branch of the root locus that starts from λ*<sup>c</sup>* might "bend" to the right (*the residue method focuses on a neighborhood of the mode to correct*)



### Low-pass Filter  $G_3$  (optional):

- in a thermal power plant, the turbine stages, the generator and the exciter are mounted on a relatively long shaft. The latter has torsional oscillation frequencies in the range 10 − 15 Hz and higher
- the PSS must not excite those frequencies
- $\bullet$  the risk is higher for a PSS using the rotor speed as input signal
- $\bullet$  in this case,  $G_3$  is a low-pass filter so that the PSS contribution is negligible at the lowest torsional frequency and above.





### **3.2 Power System Stabilizer**







### <span id="page-56-0"></span>**Transient stability:**

- depends on operating point and system parameters
- depends also on the disturbance
	- the system may be stable for disturbance 1 but not disturbance 2
	- if so, the system is insecure for 2, but as long as 2 does not happen, it can operate
	- usually, an N-1 security is required

### **Small-disturbance angle stability:**

- depends on operating point and system parameters
- does not depend on the disturbance (assumed infinitesimal and arbitrary)
- is a necessary condition for operating a power system (small disturbances are **always** present)







# <span id="page-58-0"></span>**[Introduction](#page-3-0)**

- **[Voltage Stability](#page-11-0)**
- **[Rotor Angle Stability](#page-25-0)**





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- [2] M. J. Gibbard, P. Pourbeik, and D. J. Vowles, "Small-signal stability, control and dynamic performance of power systems", University of Adelaide Press, Adelaide, 2015.